ON Semiconductor

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2-in-1 PFC and Inverter Intelligent Power Module (IPM), 600 V, 15 A

The STK57FU394AG-E is a fully-integrated PFC and inverter power stage consisting of a high-voltage driver, six motor drive IGBT's, one PFC IGBT, one PFC rectifier and a thermistor, suitable for driving permanent magnet synchronous (PMSM) motors, brushless-DC (BLDC) motors and AC asynchronous motors.

The IGBT's are configured in a 3-phase bridge with separate emitter connections for the lower legs for maximum flexibility in the choice of control algorithm.

An internal comparator and reference connected to the over-current protection circuit allows the designer to set individual over-current protection levels for the PFC and the inverter stages. Additionally, the power stage has a full range of protection functions including cross-conduction protection, external shutdown and under-voltage lockout functions.

Features

- Simple thermal design with PFC and inverter stage in one package.
- PFC operating frequency up to 40kHz
- Cross-conduction protection
- Adjustable over-current protection level
- Integrated bootstrap diodes and resistors

Certification

UL1557 (File Number : E339285)

Typical Applications

- Heat Pumps
- Home Appliances
- Industrial Fans
- Industrial Pumps

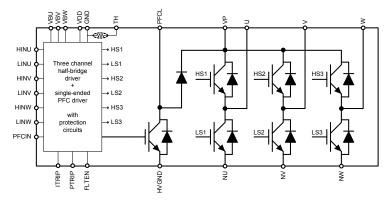


Figure 1. Functional Diagram

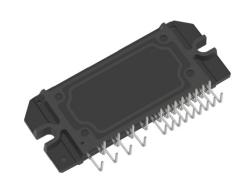
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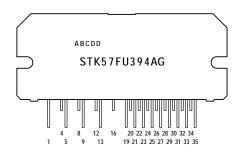
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PACKAGE PICTURE



SIP35 56x25.8 / SIP2A-3

MARKING DIAGRAM



STK57FU394AG = Specific Device Code

A = Year

B = Month

C = Production Site

DD = Factory Lot Code

Device marking is on package top side

ORDERING INFORMATION

Device	Package	Shipping (Qty / Packing)
STK57FU394AG-E	SIP35 56x25.8 / SIP2A-3 (Pb-Free)	8 / Tube

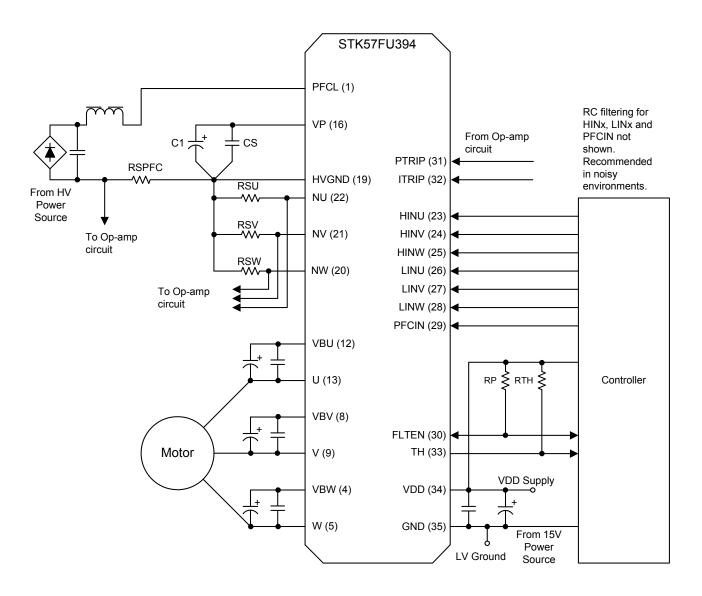


Figure 2. Application Schematic

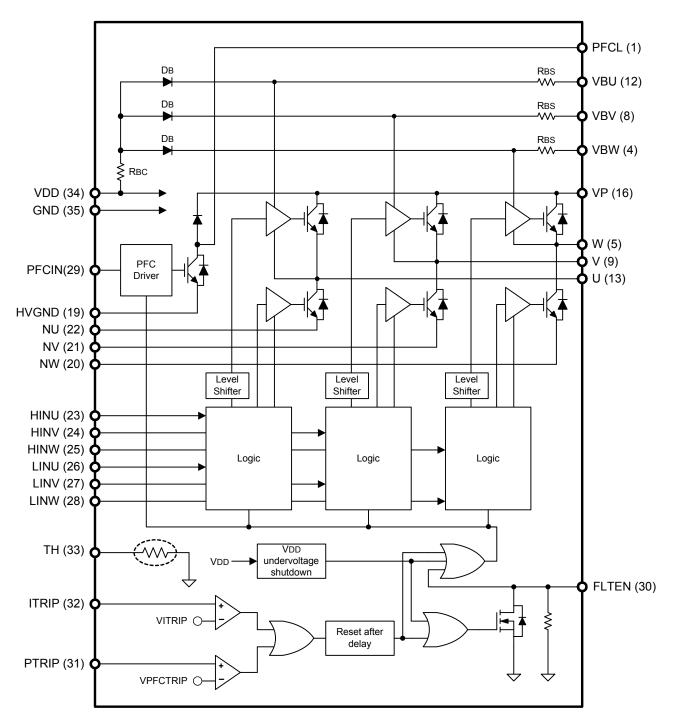


Figure 3. Simplified Block Diagram

PIN FUNCTION DESCRIPTION

Pin	Name	Description
1	PFCL	PFC Inductor Connection to IGBT and Rectifier node
4	VBW	High Side Floating Supply voltage for W phase
5	W	V phase output. Internally connected to W phase high side driver ground
8	VBV	High Side Floating Supply voltage for V phase
9	V	V phase output. Internally connected to V phase high side driver ground
12	VBU	High Side Floating Supply voltage for U phase
13	U	U phase output. Internally connected to U phase high side driver ground
16	VP	Positive PFC Output Voltage
19	HVGND	Negative PFC Output Voltage
20	NW	Low Side Emitter Connection - Phase W
21	NV	Low Side Emitter Connection - Phase V
22	NU	Low Side Emitter Connection - Phase U
23	HINU	Logic Input High Side Gate Driver - Phase U
24	HINV	Logic Input High Side Gate Driver - Phase V
25	HINW	Logic Input High Side Gate Driver - Phase W
26	LINU	Logic Input Low Side Gate Driver - Phase U
27	LINV	Logic Input Low Side Gate Driver - Phase V
28	LINW	Logic Input Low Side Gate Driver – Phase W
29	PFCIN	Logic Input PFC Gate Driver
30	FLTEN	Bidirectional FAULT output and ENABLE input
31	PTRIP	Current protection pin for PFC
32	ITRIP	Current protection pin for inverter
33	TH	Thermistor output
34	VDD	+15V Main Supply
35	GND	Negative Main Supply

Note: Pins 2, 3, 6, 7, 10, 11, 14, 15, 17 and 18 are not present

ABSOLUTE MAXIMUM RATINGS (Notes 1, 2)

Tc=25°C unless otherwise noted.

Rating		Symbol	Conditions	Value	Unit
PFC Sec	ction				
	Collector-emitter voltage	VCE	PFCL to HVGND	600	V
Repetitive peak collector current		ICP	Duty cycle 10%, pulse width 1ms	72	Α
PFC IGBT	Collector ourrent	10		36	Α
.02.	Collector current	IC	Tc=100°C	18	Α
	Maximum power dissipation	PC		73	W
	Diode reverse voltage	VRM	VP to PFCL	600	V
	Repetitive peak forward current	IFP1	Duty cycle 10%, pulse width 1ms	60	А
PFC Diode	Die de fermand ausmant	154		30	Α
Diode	Diode forward current	IF1	Tc=100°C	15	А
	Maximum power dissipation	PD1		56	W
Λ mti	Repetitive peak forward current	IFP2	Duty cycle 10%, pulse width 1ms	11	Α
Anti- parallel	Diode forward current	IF2		5	Α
Diode	Maximum power dissipation	PD2		10	W
Maximur	m AC input voltage	VAC	Single-phase Full-rectified	264	٧
Maximur	ximum output voltage Vo		In the Application Circuit	450	V
Input AC current (steady state)		lin (VAC=200V)		15	Arms
Inverter	Section				
Supply voltage		V _{CC}	VP to NU, NV, NW surge < 500V (Note 3)	450	V
Collector	r-emitter voltage	V _{CE} max	VP to U, V, W or U to NU, V to NV, W to NW	600	V
Outrout a		1.	VP, U, V, W, NU, NV, NW terminal current	±15	А
Output c	urrent	lo	VP, U, V, W, NU, NV, NW terminal current at Tc=100°C	±8	Α
Output p	eak current	lop	VP, U, V, W, NU, NV, NW terminal current, pulse width 1ms	±30	А
Maximur	m power dissipation	Pd	IGBT per 1 channel	35	W
Gate dri	ver section				
Gate driv	ver supply voltage	V _{BS} , V _{DD}	VBU to U, VBV to V, VBW to W, VDD to GND (Note 4)	-0.3 to +20.0	V
Input sig	nal voltage	VIN	HINU, HINV, HINW, LINU, LINV, LINW, PFCIN	-0.3 to $V_{\mbox{\scriptsize DD}}$	V
FLTEN t	erminal voltage	VFLTEN	FLTEN terminal	-0.3 to $V_{\hbox{\scriptsize DD}}$	V
ITRIP te	rminal voltage	VITRIP	ITRIP terminal	-0.3 to +10.0	V
PFCTRII	P terminal voltage	VPTRIP	PTRIP terminal	-1.5 to +2.0	V
Intellige	nt Power Module				
Junction	temperature	Tj	IGBT, FRD, Gate driver IC	150	°C
Storage	temperature	Tstg		-45 to +125	°C
Operatin	g case temperature	Тс	IPM case temperature	-30 to +100	°C
Tightenii	ng torque	MT	Case mounting screws	0.9	Nm
Isolation voltage		50Hz sine wa			

- Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device
- functionality should not be assumed, damage may occur and reliability may be affected.

 Refer to ELECTRICAL CHARACTERISTICS, RECOMMENDED OPERATING RANGES and/or APPLICATION INFORMATION for 2. Safe Operating parameters.
- 3. This surge voltage developed by the switching operation due to the wiring inductance between VP and NU, NV, NW terminals.
- VBS=VBU to U, VBV to V, VBW to W Test conditions : AC2500V, 1 s 4.

RECOMMENDED OPERATING RANGES (Note 6)

Rating	Symbol		Min	Тур	Max	Unit
Supply voltage	VCC	VP to HVGND, NU, NV, NW	0	280	400	V
Cata driver aumply voltage	V _{BS}	VBU to U, VBV to V, VBW to W	12.5	15	17.5	V
Gate driver supply voltage	V _{DD}	V _{DD} to GND (Note 6)	13.5	15	16.5	V
ON-state input voltage	VIN(ON)	HINU, HINV, HINW, LINU, LINV, LINW,	2.5	-	5.0	V
OFF-state input voltage	VIN(OFF)	PFCIN	0	-	0.3	V
PWM frequency(PFC)	fPWMp		1	-	40	kHz
PWM frequency(Inverter)	fPWMi		1	-	20	kHz
Dead time	DT	Turn-off to Turn-on (external)	1.5	-	-	μs
Allowable input pulse width	PWIN	ON and OFF	1	-	-	μs
Tightening torque		'M3' type screw	0.6	-	0.9	Nm

^{6.} Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

ELECTRICAL CHARACTERISTICS (Note 7)

Tc=25°C, V_{BIAS} (V_{BS}, V_{DD})=15V unless otherwise noted.

Parameter	eter Test Conditions		Min	Тур	Max	Unit
PFC Section					-	
Collector-emitter cut-off current	V _{CE} =600V	ICE	-	-	0.1	mA
Reverse leakage current (PFC Diode)	VR=600V	IR	-	-	0.1	mA
O-ll-stan and the standing and the stand	IC=30A, Tj=25°C	\/ = = (o.o.t)	-	1.9	2.5	V
Collector-emitter saturation voltage	IC=15A, Tj=100°C	V _{CE} (sat)	-	1.6	-	V
Diada farriard vallens (DEC Diada)	IF=30A, Tj=25°C	\/\\	-	2.0	2.6	V
Diode forward voltage (PFC Diode)	IF=15A, Tj=100°C	VF1	-	1.5	-	
Diode forward voltage (Anti-parallel Diode)	IF=5A, Tj=25°C	VF2	-	1.7	2.3	V
lunaki na kanana kanana kanana kanana	IGBT	θj-c(T)	-	-	1.7	°C/W
Junction to case thermal resistance	PFC Diode	θj-c(D)	-	-	2.2	°C/W
Switching characteristics					-	
Outliable at the c		tON	0.1	0.3	0.8	μs
Switching time	IC=30A, VP=300V, Tj=25°C	tOFF	0.1	0.4	0.9	μs
Diode reverse recovery time	7	trr	-	60	-	ns
Inverter section					-	
Collector-emitter leakage current	V _{CE} =600V	ICE	-	-	100	μΑ
Bootstrap diode reverse current	VR(DB)=600V	IR(BD)	-	-	100	μΑ
Collector to emitter esturation voltage	IC=15A, Tj=25°C		-	2.0	2.6	V
Collector to emitter saturation voltage	IC=8A, Tj=100°C	V _{CE} (SAT)	-	1.7	-	V
Diede ferward voltage	IF=15A, Tj=25°C	VF	-	2.1	2.7	V
Diode forward voltage	IF=8A, Tj=100°C	VF	-	1.7	-	V
Junction to case thermal resistance	IGBT	θj-c(T)	-	-	3.5	°C/W
Junction to case thermal resistance	FRD	θj-c(D)	-	-	7.2	°C/W
Switching time	IC = 15A, V _{CC} =300V, Tj=25°C	ton	0.1	0.5	1.0	μs
Switching time	10 - 10A, VCC-300V, 1J-20 0	t _{OFF}	0.2	0.7	1.2	μs
Turn-on switching loss		E _{ON}	-	200	-	μJ
Turn-off switching loss	IC = 15A, V _{CC} =300V, Tj=25°C	E _{OFF}	-	150	-	μJ
Total switching loss		E _{TOT}	-	350	-	μJ
Turn-on switching loss		E _{ON}	-	300	-	μJ
Turn-off switching loss	IC = 15A, V _{CC} =300V, Tj=100°C	E _{OFF}	-	200	-	μJ
Total switching loss		Етот	-	500	-	μJ
Diode reverse recovery energy	IC = 15A, V _{CC} =300V, Tj=100°C	E _{REC}	-	100	-	μJ
Diode reverse recovery time	(di/dt set by internal driver)	trr	-	200	-	ns
Reverse bias safe operating area	Ic=30A, V _{CE} =450V	RBSOA	Full Square	-		
Short circuit safe operating area	V _{CE} =400V, Tj=100°C	SCSOA	4	-	-	μs
Allowable offset voltage slew rate	U to NU, V to NV, W to NW	dv/dt	-50	-	50	V/ns

Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Driver Section		•	•	•		
	V _{BS} =15V (Note 4), per driver	ID	-	0.08	0.4	mA
Gate driver consumption current	V _{DD} =15V, total	ID	-	0.85	2.4	mA
High level Input voltage	HINU, HINV, HINW, LINU, LINV, LINW,	VIN H	2.5	-	-	V
Low level Input voltage	PFCIN to GND	VIN L	-	-	0.8	V
Logic 1 input current	VIN=+3.3V	I _{IN+}	-	100	143	μA
Logic 0 input current	VIN=0V	I _{IN-}	-	-	2	μA
Bootstrap diode forward voltage	IF=0.1A	VF(DB)	-	0.8	-	V
Destates a simultane interes	Resistor value for common boot charge line	RBC	-	22	-	Ω
Bootstrap circuit resistance	Resister values for separate boot charge lines	RBS	-	22	-	Ω
FLTEN terminal sink current	FLTEN: ON/VFAULT=0.1V	IoSD	-	2	-	mA
FLTEN clearance delay time		FLTCLR	1.3	1.65	2.0	ms
ELTENT A LI	VEN ON-state voltage	VEN(ON)	2.5	-	-	V
FLTEN Threshold	VEN OFF-state voltage	VEN(OFF)	-	-	0.8	V
ITRIP threshold voltage	ITRIP to GND	VITRIP	0.44	0.49	0.54	V
PTRIP threshold voltage	PTRIP to GND	VPTRIP	-0.37	-0.31	-0.25	V
ITRIP to shutdown propagation delay		tITRIP	490	600	850	ns
PTRIP to shutdown propagation delay		tPTRIP	440	550	800	ns
ITRIP and PTRIP blanking time		t _{ITRIPBL} t _{PFCTRIPBL}	290	350	-	ns
V _{DD} and V _{BS} supply undervoltage positive going input threshold		V _{DDUV+} V _{BSUV+}	10.5	11.1	11.7	٧
V _{DD} and V _{BS} supply undervoltage negative going input threshold		$V_{DDUV-} \ V_{BSUV-}$	10.3	10.9	11.5	٧
V _{DD} and V _{BS} supply undervoltage lockout hysteresis		$V_{DDUVH} \ V_{BSUVH}$	0.14	0.2	-	٧

^{7.} Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

TYPICAL CHARACTERISTICS PFC SECTION

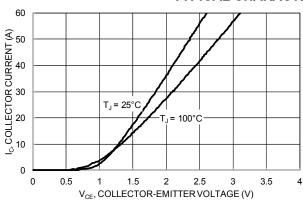


Figure 4. V_{CE} versus IC for different temperatures (V_{DD}=15V)

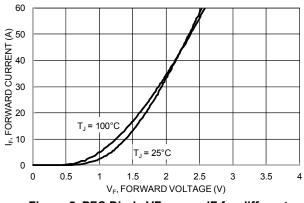


Figure 5. PFC Diode VF versus IF for different temperatures

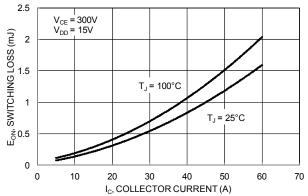


Figure 6. EON versus IC for different temperatures

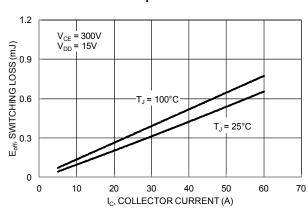


Figure 7. EOFF versus IC for different temperatures

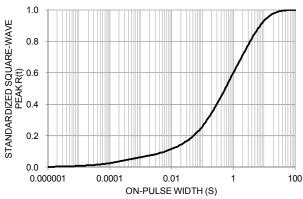


Figure 8. Thermal Impedance Plot

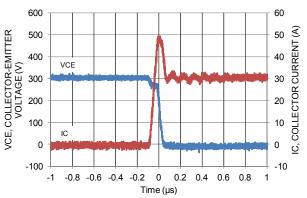


Figure 9. Turn-on waveform Tj=100°C, V_{CC}=300V

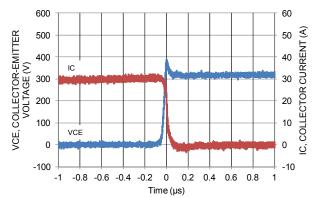


Figure 10. Turn-off waveform Tj=100°C, V_{CC}=300V

TYPICAL CHARACTERISTICS INVERTER SECTION

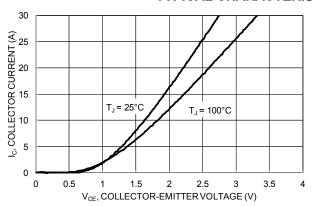


Figure 11. VCE versus ID for different temperatures $(V_{DD}=15V)$

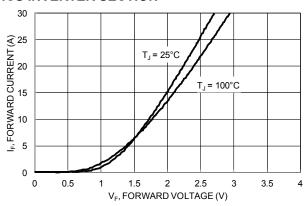


Figure 12. VF versus ID for different temperatures

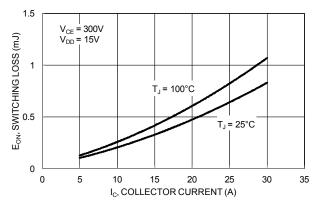


Figure 13. EON versus ID for different temperatures

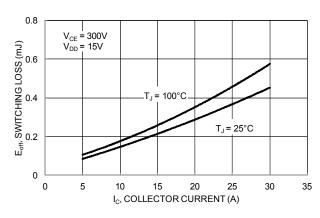


Figure 14. EOFF versus ID for different temperatures

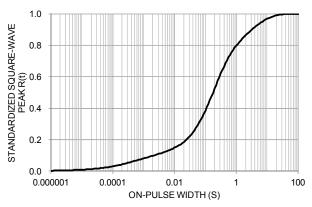


Figure 15. Thermal Impedance Plot

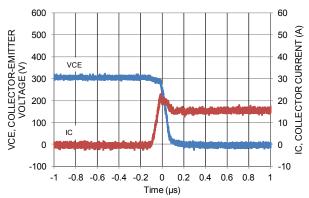


Figure 16. Turn-on waveform Tj=100°C, V_{CC}=300V

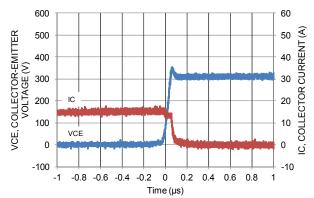


Figure 17. Turn-off waveform Tj=100°C, V_{CC}=300V

APPLICATIONS INFORMATION

Input / Output Timing Chart

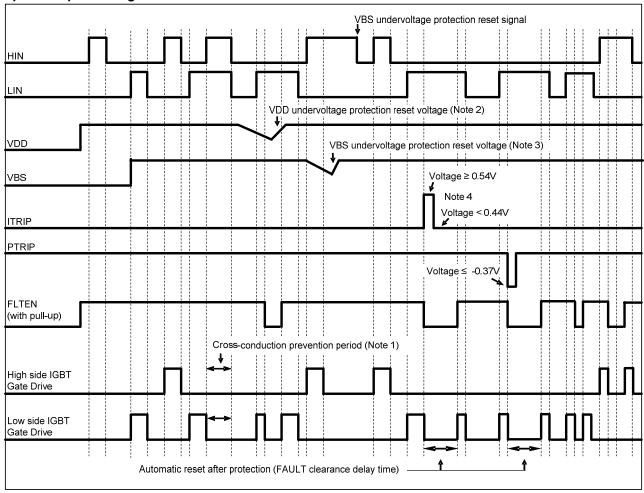


Figure 18. Input / Output Timing Chart

Notes

- 1. This section of the timing diagram shows the effect of cross-conduction prevention.
- This section of the timing diagram shows that when the voltage on V_{DD} decreases sufficiently all gate output signals will go low, switching off all six IGBTs. When the voltage on V_{DD} rises sufficiently, normal operation will resume.
- 3. This section shows that when the bootstrap voltage on VBU (VBV, VBW) drops, the corresponding high side output U (V, W) is switched off. When the voltage on VBU (VBV, VBW) rises sufficiently, normal operation will resume.
- 4. This section shows that when the voltage on ITRIP exceeds the threshold, all IGBT's are turned off. Normal operation resumes later after the over-current condition is removed. Similarly, when the voltage on PTRIP exceeds the threshold, all IGBT's are turned off. Normal operation resumes later after the over-current condition is removed
- 5. After V_{DD} has risen above the threshold to enable normal operation, the driver waits to receive an input signal on the LIN input before enabling the driver for the HIN signal.

Input / Output Logic Table

	INP	PUT			ОИТРИТ			
HIN	LIN	ITRIP	PTRIP	High side IGBT	Low side IGBT	U,V,W	FAULT	
Н	L	L	L	ON (Note 5)	OFF	VP	OFF	
L	Н	L	L	OFF	ON	NU,NV,NW	OFF	
L	L	L	L	OFF	OFF	High Impedance	OFF	
Н	Н	L	L	OFF	OFF	High Impedance	OFF	
Х	Х	Н	Х	OFF	OFF	High Impedance	ON	
Х	Х	Х	Н	OFF	OFF	High Impedance	ON	

Thermistor characteristics

Parameter	Symbol	Condition	Min	Тур	Max	Unit
Resistance	R ₂₅	Tc=25°C	99	100	101	kΩ
	R ₁₀₀	Tc=100°C	5.18	5.38	5.60	kΩ
B-Constant (25 to 50°C)	В		4208	4250	4293	K
Temperature Range			-40		+125	°C

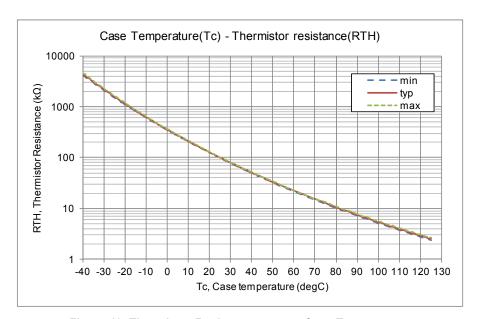


Figure 19. Thermistor Resistance versus Case Temperature

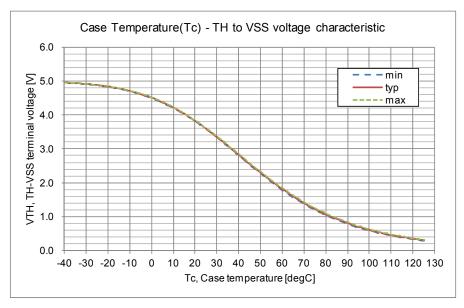


Figure 20. Thermistor Voltage versus Case Temperature Conditions: RTH=39k Ω , pull-up voltage 5.0V (see Figure 2)

Signal inputs

Each signal input has a pull-down resistor. An additional pull-down resistor of between $2.2k\Omega$ and $3.3k\Omega$ is recommended on each input to improve noise immunity.

FLTEN pin

The FLTEN pin is connected to an open-drain FAULT output requiring a pull-up resistor and an ENABLE input. If the pull-up voltage is 5V, use a pull-up resistor with a value of $6.8k\Omega$ or higher. If the pull-up voltage is 15V, use a pull-up resistor with a value of $20k\Omega$ or higher. The pulled up voltage in normal operation for the FLTEN pin should be above 2.5V, noting that it is connected to an internal ENABLE input. The FAULT output is triggered if there is a VDD undervoltage or an overcurrent condition on either the PFC or inverter stages.

Driving the FLTEN terminal pin is used to enable or shut down the built-in driver. If the voltage on the FLTEN pin rises above the positive going FLTEN threshold, the output drivers are enabled. If the voltage on the FLTEN pin falls below the negative going FLTEN threshold, the drivers are disabled.

Undervoltage protection

If VDD goes below the VDD supply undervoltage lockout falling threshold, the FAULT output is switched on. The FAULT output stays on until VDD rises above the VDD supply undervoltage lockout rising threshold. The hysteresis is approximately 200mV.

Overcurrent protection

An over-current condition is detected if the voltage on the ITRIP/PTRIP pin is larger than the reference voltage. There is a blanking time of typically 350ns to improve noise immunity. After a shutdown propagation delay of typically 0.6 us, the FAULT output is switched on.

The over-current protection threshold should be set to be equal or lower to 2 times the module rated current (Io).

An additional fuse is recommended to protect against system level or abnormal over-current fault conditions.

Capacitors on High Voltage and VDD supplies

Both the high voltage and VDD supplies require an electrolytic capacitor and an additional high frequency capacitor. The recommended value of the high frequency capacitor is between 100nF and 10 μ F.

Minimum input pulse width

When input pulse width is less than $1\mu s$, an output may not react to the pulse. (Both ON signal and OFF signal)

Calculation of bootstrap capacitor value

The bootstrap capacitor value CB is calculated using the following approach. The following parameters influence the choice of bootstrap capacitor:

- VBS: Bootstrap power supply.
 15V is recommended.
- QG: Total gate charge of IGBT at V_{BS}=15V.
 53nC
- UVLO: Falling threshold for UVLO. Specified as 12V.
- IDMAX: High side drive power dissipation. Specified as 0.4mA
- TONMAX: Maximum ON pulse width of high side IGBT.

Capacitance calculation formula:

 $CB = (QG + IDMAX * TONMAX)/(V_{BS} - UVLO)$

CB is recommended to be approximately 3 times the value calculated above. The recommended value of CB is in the range of 1 to 47 μ F, however, the value needs to be verified prior to production. When not using the bootstrap circuit, each high side driver power supply requires an external independent power supply. If the capacitors selected are 47 μ F or more, a series resistor of 20 Ω should be added in series with the three capacitors to limit the current. The resistors should be inserted between VBU and U, VBV and V and VBW and W.

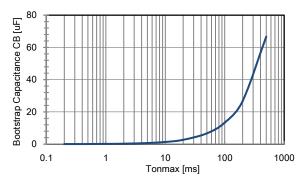


Figure 21. Bootstrap capacitance versus Tonmax

Mounting Instructions

Item	Recommended Condition
Pitch	56.0±0.1mm (Please refer to Package Outline Diagram)
Screw	diameter : M3 Screw head types: pan head, truss head, binding head
Washer	Plane washer The size is D:7mm, d:3.2mm and t:0.5mm JIS B 1256
Heat sink	Material: Aluminum or Copper Warpage (the surface that contacts IPM) : –50 to 100 μm Screw holes must be countersunk. No contamination on the heat sink surface that contacts IPM.
Torque	Temporary tightening: 20 to 30 % of final tightening on first screw Temporary tightening: 20 to 30 % of final tightening on second screw Final tightening: 0.6 to 0.9Nm on first screw Final tightening: 0.6 to 0.9Nm on second screw
Grease	Silicone grease. Thickness: 100 to 200 µm Uniformly apply silicone grease to whole back. Thermal foils are only recommended after careful evaluation. Thickness, stiffness and compressibility parameters have a strong influence on performance.

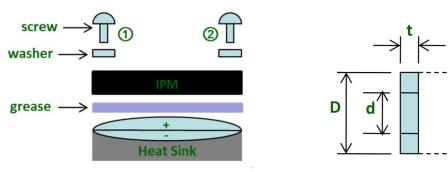


Figure 22. Mount IPM on a Heat Sink

Figure 23. Size of Washer

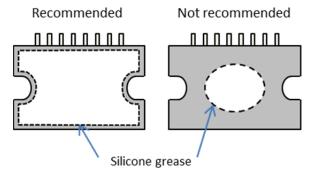


Figure 24. Uniform Application of Grease Recommended

Steps to mount an IPM on a heat sink

1st: Temporarily tighten maintaining a left/right balance.

2nd: Finally tighten maintaining a left/right balance.

TEST CIRCUITS

■ ICE, IR(DB)

	U+	V+	W+	U-	V-	W-	PFC IGBT
Α	16	16	16	13	9	5	1
В	13	9	5	22	21	20	19

U+,V+,W+ : High side phase U-,V-,W- : Low side phase

	U(DB)	V(DB)	W(DB)	PFC Diode
Α	12	8	4	16
В	35	35	35	1

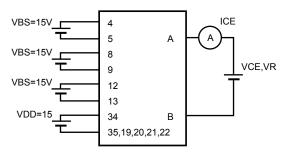


Figure 25. Test Circuit for ICE

■ VCE(sat) (Test by pulse)

	U+	V+	W+	U-	V-	W-	PFC IGBT
Α	16	16	16	13	9	5	1
В	13	9	5	22	21	20	19
С	23	24	25	26	27	28	29

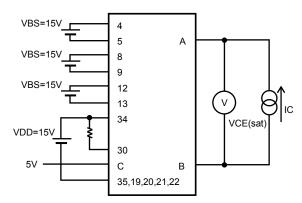


Figure 26. Test circuit for $V_{\text{CE(SAT)}}$

■ V_F (Test by pulse)

	U+	V+	W+	U-	V-	W-
Α	16	16	16	13	9	5
В	13	9	5	22	21	20

	U(DB)	V(DB)	W(DB)	PFC Diode	Anti-parallel Diode
Α	12	8	4	16	1
В	34	34	34	1	19

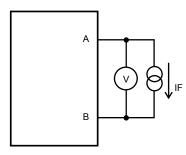


Figure 27. Test circuit for V_F

■ I_D

	VBS U+	VBS V+	VBS W+	V_{DD}
Α	12	8	4	34
В	13	9	5	35

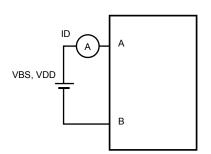
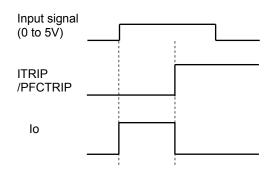


Figure 28. Test circuit for I_{D}

■ VITRIP, VPTRIP

	VITRIP(U-)	VPTRIP
Α	13	1
В	22	19
С	26	29
D	32	31



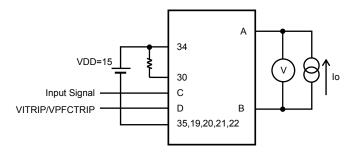
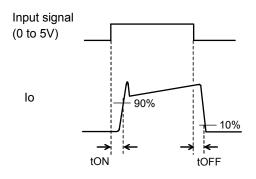


Figure 29. Test circuit for ITRIP.PTRIP

■ Switching time (The circuit is a representative example of the lower side U phase.)

	U+	V+	W+	U-	V-	W-	PFC IGBT
Α	16	16	16	16	16	16	16
В	22	21	20	22	21	20	19
С	13	9	5	13	9	5	1
D	22	21	20	16	16	16	16
E	23	24	25	26	27	28	29



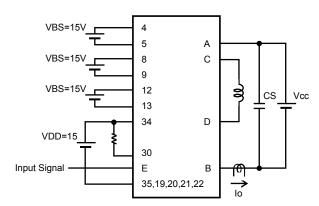


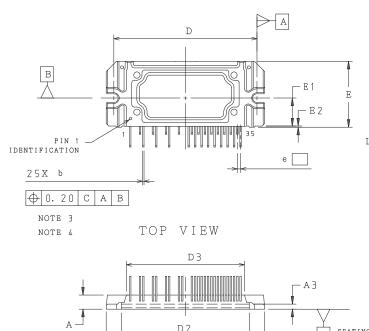
Figure 30. Test circuit for switching time

Package Dimensions

unit: mm

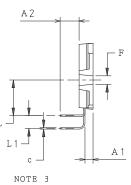
SIP35 56x25.8 / SIP2A-3

CASE 127DY **ISSUE O**



D.

SIDE VIEW



END VIEW

	MILLIMETERS				
DIM	MIN.	NOM.	MAX.		
Α	5. 00	5.50	6.00		
A 1	2. 70	3. 20	3. 70		
A 2	6. 90	7. 40	7. 90		
A 3	1. 50	2.00	2.50		
b	0. 55	0.60	0.80		
С	0.45	0.50	0.70		
D	55. 50	56.00	56, 50		
D1	61.50	62.00	62.50		
D 2	49.50	50.00	50.50		
D3	45.70	46.20	46.70		
E	25. 30	25.80	26.30		
E 1	10.90 REF				
E 2	0.00	0.50	1. 00		
е	1. 27 BSC				
F	2. 90	3. 40	3. 90		
L	13.40	13. 90	14.40		
L1	4.50	5. 00	5. 50		

SEATING

- DIMENSIONING AND TOLERANCING PER. ASME Y14.5M. 2009.
- CONTROLLING DIMENSION: MILLIMETERS
- DIMENSIONS & AND c APPLY TO THE PLATED LEAD AND ARE MEASURED BETWEEN 1.00 AND 2.00 FROM THE LEAD TIP.
- POSITION OF THE LEADS IS DETERMINED AT THE ROOT OF THE LEAD WHERE IT EXITS THE PACKAGE BODY.
- MIRROR SURFACE MARK INDICATES PIN 1 POSITION
- MISSING PINS ARB: 2. 3. 6. 7. 10. 11. 14. 15. 17. AND 18.

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