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Current Mode PWM Control Circuit with 50% Max Duty Cycle

Description

The CS3845B provides all the necessary features to implement off-line fixed frequency current-mode control with a minimum number of external components.

The CS3845B incorporates a precision temperature-controlled oscillator to minimize variations in frequency. An internal toggle flip-flop, which blanks the output off every other clock cycle, ensures that the duty-cycle never

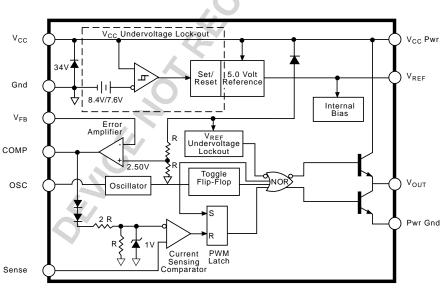
exceeds 50%. The undervoltage lockout feature ensures that V_{REF} is stabilized within specification before the output stage is enabled. The CS3845B has been optimized for lower start up current (500µA max).

Other features include 1% trimmed band gap reference, pulse-by-pulse current limiting, and a high-current totem pole output for driving capacitive loads.

Absolute Maximum

Supply Voltage (I _{CC} <30mA)	Self Limiting
Supply Voltage (Low Impedance Source)	30V
Output Current	
Output Energy (Capacitive Load)	5µJ
Analog Inputs (V _{FB} , V _{SENSE})	–0.3V to 5.5V
Error Amp Output Sink Current	10mA
Lead Temperature Soldering	
	00000 1

Wave Solder (through hole styles only)......10 sec. max, 260°C peak Reflow (SMD styles only)......60 sec. max above 183°C, 230°C peak



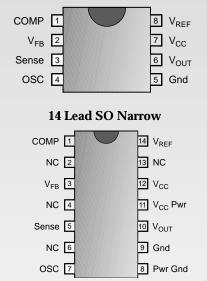
Block Diagram

Features

- Optimized for Off-line Use Temperature
- Compensated Oscillator
- 50% Maximum Duty-cycle Clamp
- Low Start-up Current (500µA max)
- Pulse-by-pulse Current Limiting
- Undervoltage Active Pull Down
- **Double Pulse Suppression**
- 1% Trimmed Bandgap Reference
- High Current Totem Pole Output

Package Options

8 Lead PDIP & SO

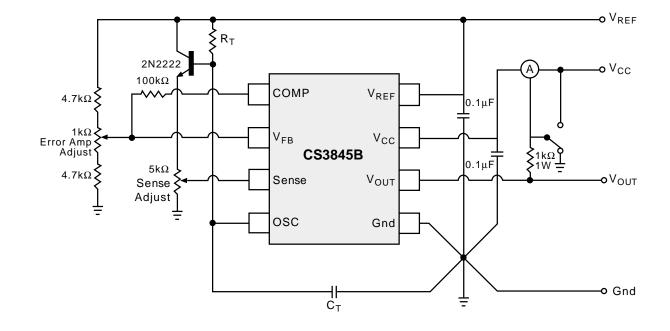


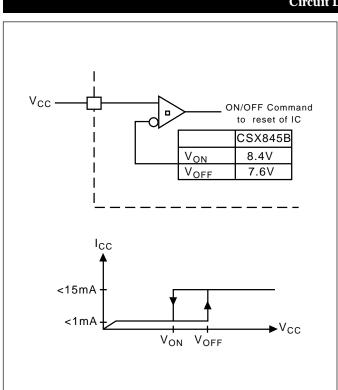
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PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
Reference Section					
Output Voltage	$T_J = 25^{\circ}C$, $I_{REF}=1mA$	4.90	5.00	5.10	V
Line Regulation	$12 \leq V_{CC} \leq 25V$		6	20	mV
Load Regulation	$1 \leq I_{REF} \leq 20 mA$		6	25	mV
Temperature Stability	(Note 1)		0.2	0.4	mV/°
Total Output Variation	Line, Load, Temp. (Note 1)	4.82		5.18	V
Output Noise Voltage	$10Hz \leq f \leq 10kHz, T_J$ = 25°C (Note 1)		50		μV
Long Term Stability	T _A = 125°C, 1000 Hrs. (Note 1)		5	25	mV
Output Short Circuit	$T_A = 25^{\circ}C$	-30	-100	-180	mA
Oscillator Section					
Initial Accuracy	Sawtooth Mode, $T_I = 25^{\circ}C$	47	52	57	kHz
Voltage Stability	$12 \le V_{CC} \le 25V$		0.2	1.0	%
Temperature Stability	Sawtooth Mode $T_{MIN} \le T_A \le T_{MAX}$				
	(Note 1)		5		%
Amplitude	V _{OSC} (peak to peak)(Note 1)		1.7		V
Discharge	$\begin{array}{l} T_{J} = 25^{\circ}C; \mbox{ (Note 1)} \\ T_{MIN} \leq T_{A} \leq T_{MAX} \mbox{ (Note 1)} \end{array}$	7.5 7.2	8.3	9.3 9.5	mA mA
Error Amp Section		0.40		0.50	V
Input Voltage	$V_{\rm COMP} = 2.5 V$	2.42	2.50	2.58	
Input Bias Current	$V_{FB} = 0V$	05	-0.3	-2.0	μA
AVOL	$2 \le V_{OUT} \le 4V$	65	90		dB
Unity Gain Bandwidth	(Note 1)	0.7	1.0		MHz
PSRR	$12 \leq V_{CC} \leq 25V$	60	70		dB
Output Sink Current	$V_{FB} = 2.7V, V_{COMP} = 1.1V$	2	6		mA
Output Source Current	$V_{FB} = 2.3V$, $V_{COMP} = 5V$	-0.5	-0.8		mA
V _{OUT} HIGH	$V_{FB} = 2.3V$, $R_L 15k\Omega$ to Gnd	5	6		V
V _{OUT} LOW	V_{FB} = 2.7V, R_L = 15k Ω to V_{REF}		0.7	1.1	V
Current Sense Section					
Gain	(Notes 2&3)	2.85	3.00	3.15	V/V
Maximum Input Signal	$V_{\text{COMP}} = 5V$ (Note 2)	0.9	1.0	1.1	V
PSRR	$12 \leq V_{CC} \leq 25V$ (Note 2)		70		dB
Input Bias Current	$V_{Sense} = 0V$		-2	-10	μA
Delay to Output	<u>$T_{J} = 25^{\circ}C$ (Note 1)</u>		150	300	ns
Output Section					
Output Low Level	$I_{SINK} = 20 \text{mA}$		0.1	0.4	V
-	$I_{SINK} = 200 \text{mA}$		1.5	2.2	V
Output High Level	$I_{\text{SOURCE}} = 20 \text{mA}$	13.0	13.5		V
	$I_{\text{SOURCE}} = 200 \text{mA}$	12.0	13.5		V
Rise Time	$T_J = 25^{\circ}C, C_L = 1nF$ (Note 1)		50	150	ns
Fall Time	$T_J = 25^{\circ}C, C_L = 1nF$ (Note 1)		50	150	ns

Electrical Characteristics: continued PARAMETER TEST CONDITIONS MIN TYP MAX UNITS						
PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNITS	
Total Standby Current						
Start-Up Current			300	500	μΑ	_
Operating Supply Current	$V_{FB} = V_{Sense} = 0V R_T = 10k\Omega, C_T = 3.3nH$	1	11	17	mA	
V _{CC} Zener Voltage	I _{CC} =25mA		34		V	
PWM Section Maximum Duty Cycle Minimum Duty Cycle		46	48	50 0	%	
Under-Voltage Lockout Sect	ion					
Start Threshold		7.8	8.4	9.0	V	
Min. Operating Voltage	After Turn On	7.0	7.6	8.2	V	
 Notes: 1. These parameters, althouged in production. 2. Parameter measured at transmission 	h guaranteed, are not 100% test- 3. Gai p point of latch with V _{FB} =0	n defined as: A	$\Delta = \frac{\Delta V_{COMP}}{\Delta V_{Sense}};$	$0 \le V_{Sense} \le 0.8$	V.	

Package Pin Description			
PACKAGE PIN #		PIN SYMBOL	FUNCTION
8L PDIP & SO	14L SO		
1	1	СОМР	Error amp output, used to compensate error amplifier
2	3	V_{FB}	Error amp inverting input
3	5	Sense	Noninverting input in Current Sense Comparator
4	7	OSC	Oscillator timing network with Capacitor to Gnd, resistor to V_{REF}
5	9	Gnd	Ground
6	10	V _{OUT}	Output drive pin
7	12	V _{CC}	Positive power supply
8	14	V_{REF}	Output of 5V internal reference
	8	Pwr Gnd	Output driver Gnd
	11	V _{CC} Pwr	Output driver positive supply
	2, 4, 6, 13	NC	No Connection





Circuit Description

Undervoltage Lockout

During Undervoltage Lockout (Figure 1), the output driver is biased to sink minor amounts of current. The output should be shunted to ground with a resistor to prevent activating the power switch with extraneous leakage currents.

PWM Waveform

To generate the PWM waveform, the control voltage from the error amplifier is compared to a current sense signal which represents the peak output inductor current (Figure 2). An increase in V_{CC} causes the inductor current slope to increase, thus reducing the duty cycle. This is an inherent feed-forward characteristic of current mode control, since the control voltage does not have to change during changes of input supply voltage.

When the power supply sees a sudden large output current increase, the control voltage will increase allowing the duty cycle to momentarily increase. Since the duty cycle tends to exceed the maximum allowed to prevent transformer saturation in some power supplies, the internal oscillator waveform provides the maximum duty cycle clamp as programmed by the selection of OSC components.

Figure 1: Startup voltage for CS3845B.

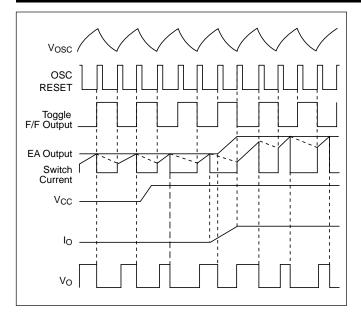


Figure 2: Timing Diagram

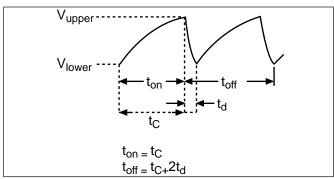


Figure 3: Timing Parameters.

Setting the Oscillator

The parameters T_c and T_d can be determined as follows:

$$\begin{split} t_{c} &= R_{T}C_{T} \ln \left(\frac{V_{REF} - V_{lower}}{V_{REF} - V_{upper}} \right) \\ t_{d} &= R_{T}C_{T} \ln \left(\frac{V_{REF} - I_{d}R_{T} - V_{lower}}{V_{REF} - I_{d}R_{T} - V_{upper}} \right) \end{split}$$

Substituting in typical values for the parameters in the above formulas:

 $V_{REF}=5.0V,\,V_{upper}=2.7V,\,V_{lower}=1.0V,\,I_d=8.3mA,$ then

 $t_c \approx 0.5534 R_T C_T$

$$t_{d} = R_{T}C_{T} \ln \left(\frac{2.3 - 0.0083 R_{T}}{4.0 - 0.0083 R_{T}}\right)$$

For better accuracy R_T should be $\geq 10k\Omega$.

Grounding

High peak currents associated with capacitive loads necessitate careful grounding techniques. Timing and bypass capacitors should be connected close to Gnd in a single point ground.

The transistor and $5k\Omega$ potentiometer are used to sample the oscillator waveform and apply an adjustable ramp to Sense.

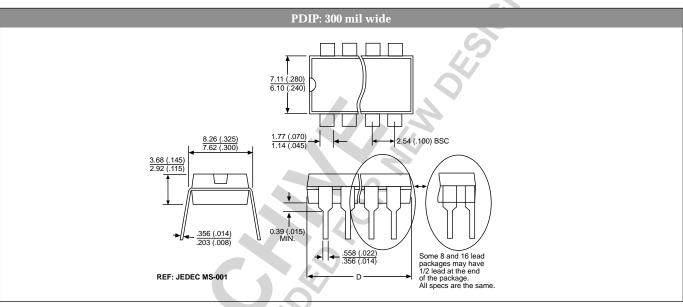
CS3845B

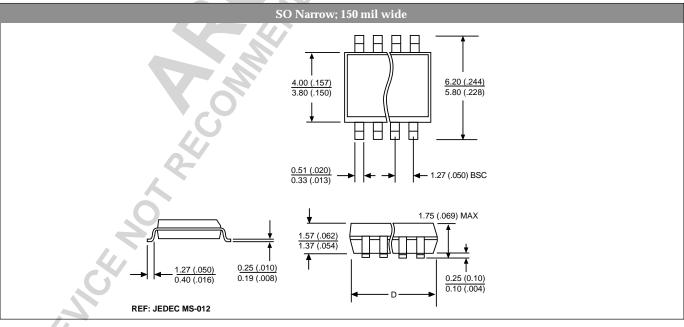
Package Specification

PACKAGE DIMENSIONS IN mm (INCHES)

	D			
Lead Count	Met	ric	Eng	lish
	Max	Min	Max	Min
8 Lead PDIP	10.16	9.02	.400	.355
8 Lead SO Narrow	5.00	4.80	.197	.189
14 Lead SO Narrow	8.75	8.55	.344	.337

PACKAGE THERMAL DATA					
Therma	l Data	8 L	8L	14 L	
		PDIP	SO	SO	
R _{QJC}	typ	52	45	30	°C/W
R _{QJA}	typ	100	165	125	°C/W





Ordering	Information

Part N. aber	Description
CS3845BGN8	8L PDIP
CS3845BGD8	8L SO
CS3845BGDR8	8L SO (tape & reel)
CS3845BGD14	14L SO
CS3845BGDR14	14L SO (tape & reel)

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