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Silicon Carbide (SiC) Cascode JFET – EliteSiC, Power N-Channel, TO-263-7, 1700 V, 410 mohm

UF3C170400B7S

Description

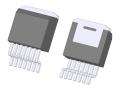
This SiC FET device is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gatedrive characteristics allows for a true "drop-in replacement" to Si IGBTs, Si FETs, SiC MOSFETs or Si superjunction devices. Available in the TO-263-7 package, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads, and any application requiring standard gate drive.

Features

- On-resistance $R_{DS(on)}$: 410 m Ω (Typ)
- Operating Temperature: 175 °C (Max)
- Excellent Reverse Recovery: $Q_{rr} = 70 \text{ nC}$
- Low Body Diode V_{FSD}: 1.5 V
- Low Gate Charge: $Q_G = 23.1 \text{ nC}$
- Low Intrinsic Capacitance
- ESD Protected: HBM Class 2 and CDM Class C3
- This Device is Halogen Free and RoHS Compliant with Exemption 7a, Pb-Free 2LI (on second level interconnection)

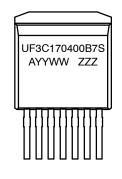
Typical Applications

- Switching Power Supplies
- Auxiliary Power Supplies
- Load Switches



TO-263-7 CASE 418BA

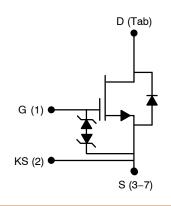
MARKING DIAGRAM



UF3C170400B7S = Specific Device Code A = Assembly Location

А	= Assembly Loca
YY	= Year
WW	= Work Week
ZZZ	= Lot ID

PIN CONNECTIONS



ORDERING INFORMATION

See detailed ordering and shipping information on page 9 of this data sheet.

MAXIMUM RATINGS

Parameter	Symbol	Test Conditions	Value	Unit
Drain-source Voltage	V _{DS}		1700	V
Gate-source Voltage	V _{GS}	DC	–25 to +25	V
Continuous Drain Current (Note 1)	I _D	T _C = 25 °C	7.6	А
		T _C = 100 °C	5.9	А
Pulsed Drain Current (Note 2)	I _{DM}	T _C = 25 °C	14	А
Single Pulsed Avalanche Energy (Note 3)	E _{AS}	L = 15 mH, I _{AS} = 1.25 A	11.7	mJ
Power Dissipation	P _{tot}	T _C = 25 °C	100	W
Maximum Junction Temperature	T _{J, max}		175	°C
Operating and Storage Temperature	T _J , T _{STG}		–55 to 175	°C
Reflow Soldering Temperature	T _{solder}	Reflow MSL 1	260	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.
1. Limited by T_{J, max}.
2. Pulse width t_p limited by T_{J, max}.
3. Starting T_J = 25 °C.

THERMAL CHARACTERISTICS

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Thermal Resistance, Junction-to-Case	$R_{\theta JC}$		-	1.2	1.5	°C/W

ELECTRICAL CHARACTERISTICS (T_J = +25 °C unless otherwise specified)

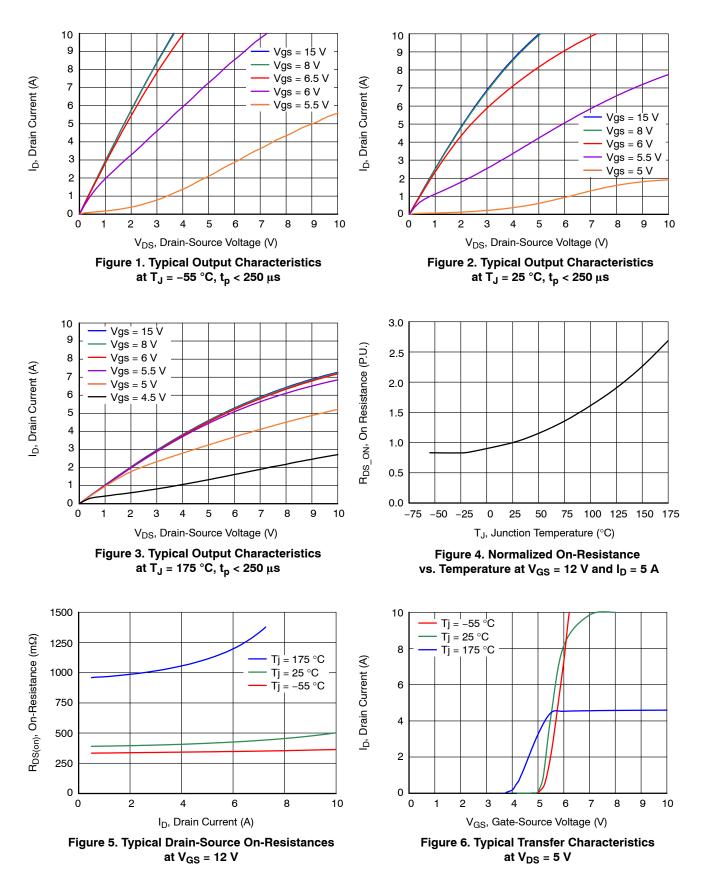
Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
TYPICAL PERFORMANCE – STATIC					4	
Drain-source Breakdown Voltage	BV _{DS}	V _{GS} = 0 V, I _D = 1 mA	1700	-	_	V
Total Drain Leakage Current	I _{DSS}	V _{DS} = 1700 V, V _{GS} = 0 V, T _J = 25 °C	-	1.5	60	μA
		V _{DS} = 1700 V, V _{GS} = 0 V, T _J = 175°C	-	5.5	_	
Total Gate Leakage Current	I _{GSS}	V _{DS} = 0 V, T _J = 25 °C, V _{GS} = -20 V / +20 V	-	6	±20	μA
Drain-source On-resistance	R _{DS(on)}	V_{GS} = 12 V, I _D = 5 A, T _J = 25°C	-	410	515	mΩ
		V _{GS} = 12 V, I _D = 5 A, T _J = 125°C	_	780	_	
		V _{GS} = 12 V, I _D = 5 A, T _J = 175°C	_	1070	_	
Gate Threshold Voltage	V _{G(th)}	V _{DS} = 5 V, I _D = 10 mA	3	4.7	6	V
Gate Resistance	R _G	f = 1 MHz, open drain	-	4.1	_	Ω
TYPICAL PERFORMANCE – REVERSE DIOD	E					
Diode Continuous Forward Current (Note 1)	Is	T _C = 25 °C	-	-	7.6	А
Diode Pulse Current (Note 2)	I _{S, pulse}	T _C = 25 °C	-	_	14	А
Forward Voltage	V _{FSD}	V _{GS} = 0 V, I _S = 2 A, T _J = 25 °C	-	1.5	1.75	V
J.		V _{GS} = 0 V, I _S = 2 A, T _J = 175 °C	_	2.4	_	
Reverse Recovery Charge	Q _{rr}	V _{DS} = 1200 V, I _S = 5 A, V _{GS} = -5 V,	_	70	_	nC
Reverse Recovery Time	t _{rr}	R _{G_EXT} = 10 Ω, di/dt = 4000 A/μs, T _{.1} = 25 °C	-	29	_	ns
Reverse Recovery Charge	Q _{rr}	V _{DS} = 1200 V, I _S = 5 A, V _{GS} = -5 V,	_	67	_	nC
Reverse Recovery Time	t _{rr}	R _{G_EXT} = 10 Ω, di/dt = 4000 A/μs, T _{.I} = 150 °C	-	27	_	ns
TYPICAL PERFORMANCE – DYNAMIC		.1 = 100 0				
Input Capacitance	C _{iss}	V _{DS} = 1200 V, V _{GS} = 0 V,	_	734	_	pF
Output Capacitance	C _{oss}	f = 100 kHz	_	13.6	_	р. Р.
Reverse Transfer Capacitance	C _{rss}	-	_	2	_	
Effective Output Capacitance, Energy Related	C _{oss(er)}	V _{DS} = 0 V to 1200 V, V _{GS} = 0 V	_	15.5	_	pF
Effective Output Capacitance, Time Related	C _{oss(tr)}		_	28	_	
C _{OSS} Stored Energy	E _{oss}	V _{DS} = 1200 V, V _{GS} = 0 V	_	11.2	_	μJ
Total Gate Charge	Q _G	V _{DS} = 1200 V, I _D = 5 A,	_	23.1	_	nC
Gate-drain Charge	Q _{GD}	$V_{GS} = 0 V$ to 15 V	_	6.5	_	
Gate-source Charge	Q _{GS}		_	5.6	_	
Turn-on Delay Time	t _{d(on)}	V _{DS} = 1200 V, I _D = 5 A,	_	43	_	ns
Rise Time	t _r	Gate Driver = 0 V to +15 V, R _{G. EXT} = 50 Ω,	-	16	-	
Turn-off Delay Time	t _{d(off)}	Inductive Load,	-	102	_	
Fall Time	t _f	FWD: 2x UJ3D1202TS in series, T _J = 25 °C	-	27.5	_	
Turn-on Energy	E _{ON}	1	-	143	_	μJ
Turn-off Energy	E _{OFF}	1	-	29	_	
Total Switching Energy	E _{TOTAL}	1	-	172	-	1

ELECTRICAL CHARACTERISTICS (T_J = +25 $^{\circ}$ C unless otherwise specified) (continued)

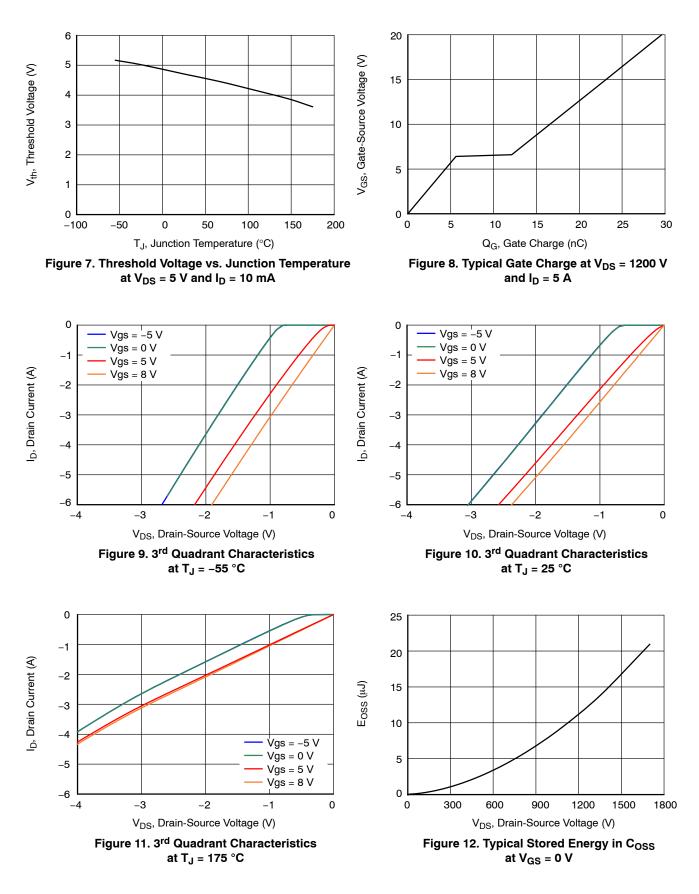
Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit		
TYPICAL PERFORMANCE - DYNAMIC								
Turn-on Delay Time	t _{d(on)}	$V_{DS} = 1200 \text{ V}, \text{ I}_{D} = 5 \text{ A},$	-	33	-	ns		
Rise Time	tr	Gate Driver = 0 V to +15 V, R _{G. EXT} = 50 Ω,	-	15	-			
Turn-off Delay Time	t _{d(off)}	Inductive Load, FWD: 2x UJ3D1202TS in series, $T_J = 150 \ ^{\circ}C$	-	99	-			
Fall Time	t _f		-	35	-			
Turn-on Energy	E _{ON}		-	127	-	μJ		
Turn-off Energy	E _{OFF}		-	27	-			
Total Switching Energy	E _{TOTAL}		_	154	-			

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

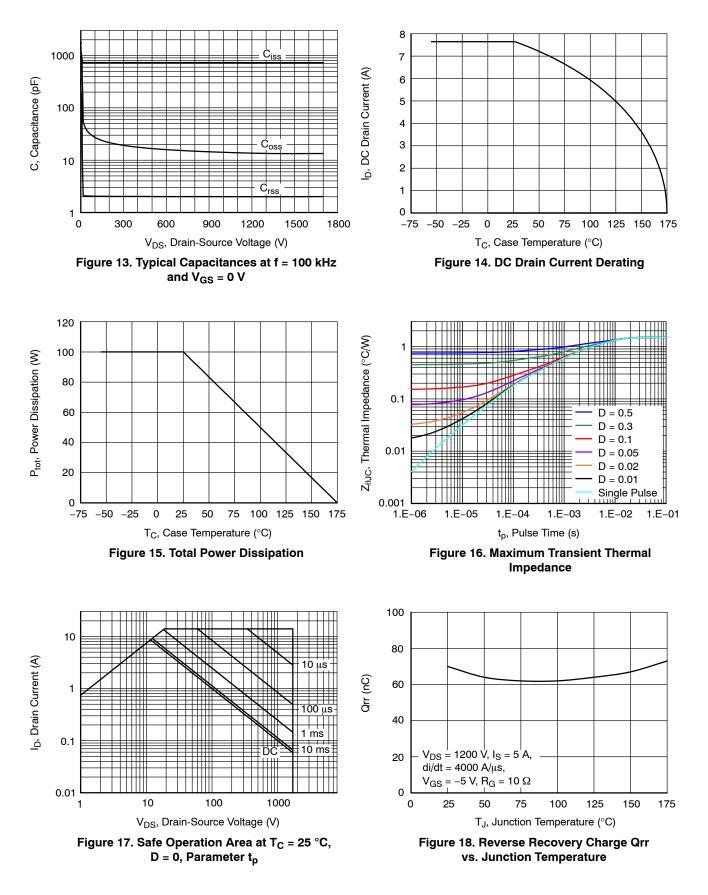
TYPICAL PERFORMANCE DIAGRAMS



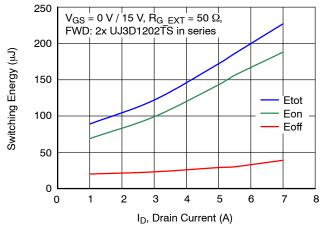
TYPICAL PERFORMANCE DIAGRAMS (continued)

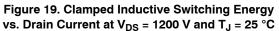


TYPICAL PERFORMANCE DIAGRAMS (continued)



TYPICAL PERFORMANCE DIAGRAMS (continued)





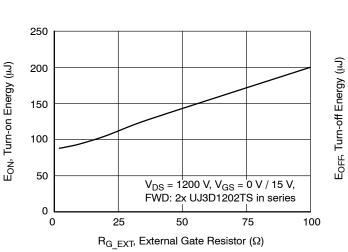


Figure 21. Clamped Inductive Switching Turn-on Energy vs. Gate Resistor R_{G_EXT}

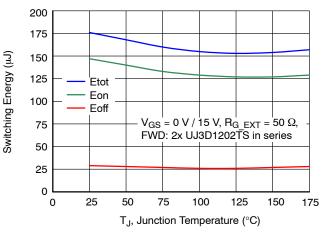


Figure 20. Clamped Inductive Switching Energy vs. Junction Temperature at V_{DS} = 1200 V, and I_D = 5 A

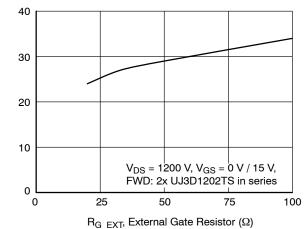


Figure 22. Clamped Inductive Switching Turn-off Energy vs. Gate Resistor $R_{G\ EXT}$

APPLICATIONS INFORMATION

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.onsemi.com.

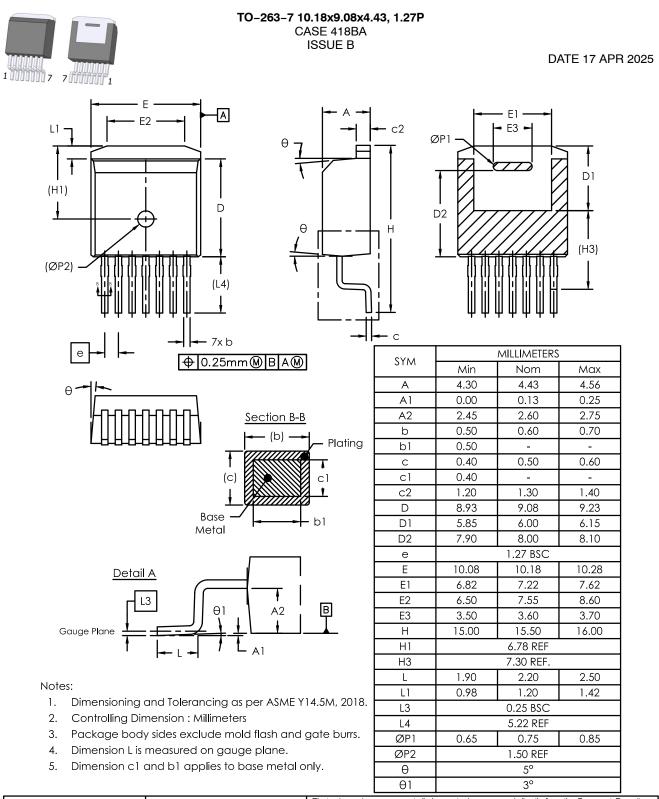
A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the **onsemi** website at <u>www.onsemi.com</u>

ORDERING INFORMATION

Part Number	Marking	Package	Shipping [†]
UF3C170400B7S	UF3C170400B7S	TO-263-7	800 / Tape & Reel

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, <u>BRD8011/D</u>.

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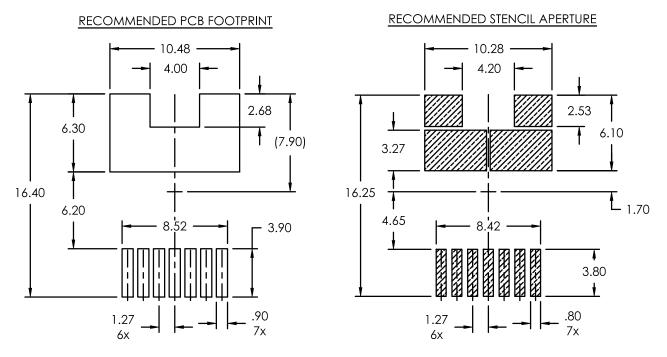


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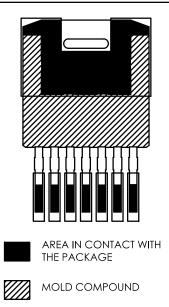
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DATE 17 APR 2025



NOTE: LAND PATTERN AND STENCIL APERTURE DIMENSIONS SERVE ONLY AS AN INITIAL GUIDE. END-USER PCB DESIGN RULES AND TOLERANCES SHOULD ALWAYS PREVAIL.



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