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Silicon Carbide (SiC) Cascode JFET - EliteSiC, Power N-Channel, TO247-3, 1200 V, 72 mohm

UF4C120070K3S

Description

The UF4C120070K3S is a 1200 V, 72 m Ω G4 SiC FET. It is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows for a true "drop-in replacement" to Si IGBTs, Si FETs, SiC MOSFETs or Si super-junction devices. Available in the TO247-3 package, this device exhibits ultra-low gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads and any application requiring standard gate drive.

Features

- On-resistance R_{DS(on)}: 72 mΩ
- Maximum Operating Temperature of 175 °C
- Excellent Reverse Recovery: $Q_{rr} = 72 \text{ nC}$
- Low Body Diode V_{FSD}: 1.43 V
- Low Gate Charge: $Q_G = 37.8 \text{ nC}$
- Threshold Voltage V_{G(th)}: 4.8 V (typ) Allowing 0 to 15 V Drive
- Low Intrinsic Capacitance
- ESD Protected: HBM Class 2 and CDM Class C3
- This Device is Pb-Free, Halogen Free and is RoHS Compliant

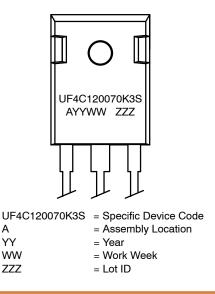
Typical Applications

- EV Charging
- PV Inverters
- Switch Mode Power Supplies
- Power Factor Correction Modules
- Motor Drives
- Induction Heating

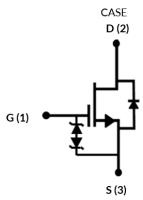


TO247-3 15.90x20.96x5.03, 5.44P CASE 340AK

MARKING DIAGRAM



PIN CONNECTIONS



ORDERING INFORMATION

See detailed ordering and shipping information on page 9 of this data sheet.

MAXIMUM RATINGS

Parameter	Symbol	Test Conditions	Value	Unit
Drain-source Voltage	V _{DS}		1200	V
Gate-source Voltage	V _{GS}	DC	-20 to +20	V
		AC (f > 1 Hz)	-25 to +25	
Continuous Drain Current (Note 1)	I _D	T _C = 25 °C	27.5	А
		T _C = 100 °C	20.7	А
Pulsed Drain Current (Note 2)	I _{DM}	T _C = 25 °C	83	А
Single Pulsed Avalanche Energy (Note 3)	E _{AS}	L = 15 mH, I _{AS} = 2.2 A	36	mJ
SiC FET dv/dt Ruggedness	dv/dt	V _{DS} ≤ 800 V	200	V/ns
Power Dissipation	P _{tot}	T _C = 25 °C	217	W
Maximum Junction Temperature	T _{J,max}		175	°C
Operating and Storage Temperature	T _J , T _{STG}		–55 to 175	°C
Max. Lead Temperature for Soldering, 1/8" from Case for 5 Seconds	TL		250	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected. 1. Limited by $T_{J,max}$ 2. Pulse width t_p limited by $T_{J,max}$ 3. Starting $T_J = 25 \text{ °C}$

THERMAL CHARACTERISTICS

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Thermal Resistance, Junction-to-Case	R_{\thetaJC}		-	0.53	0.69	°C/W

ELECTRICAL CHARACTERISTICS (T_J = +25 °C unless otherwise specified)

Parameter	Symbol	Test Conditions		Min	Тур	Max	Unit
TYPICAL PERFORMANCE – STATIC							
Drain-source Breakdown Voltage	BV _{DS}	V _{GS} = 0 V, I _D = 1 mA		1200	-	-	V
Total Drain Leakage Current	I _{DSS}	V_{DS} = 1200 V, V_{GS} = 0 V, T_{J} = 25 $^{\circ}C$		-	0.4	18	μΑ
		V_{DS} = 1200 V, V_{GS} = 0 V, T_{J} = 175°C		-	10	-	
Total Gate Leakage Current	I _{GSS}	V_{DS} = 0 V, T _J = 25 °C, V_{GS} = -20 V/ +20 V		-	6	20	μΑ
Drain-source On-resistance	R _{DS(on)}	V_{GS} = 12 V, I _D = 20 A T _J = 25 °C		-	72	91	mΩ
			T _J = 125 °C	-	140	-	
			T _J = 175 °C	-	197	-	
Gate Threshold Voltage	V _{G(th)}	V _{DS} = 5 V, I _D = 10 mA		4	4.8	6	V
Gate Resistance	R _G	f = 1 MHz, open drain		-	4.5	-	Ω

TYPICAL PERFORMANCE - REVERSE DIODE

Diode Continuous Forward Current (Note 4)	ا _S	T _C = 25 °C	-	-	27.5	A
Diode Pulse Current (Note 5)	I _{S,pulse}	T _C = 25 °C	-	-	83	А
Forward Voltage	V _{FSD}	V_{GS} = 0 V, I _S = 10 A, T _J = 25 °C	-	1.43	1.64	V
		V_{GS} = 0 V, I _S = 10 A, T _J = 175 °C	-	2.38	-	
Reverse Recovery Charge	Q _{rr}	$V_{DS} = 800 \text{ V}, \text{ I}_{S} = 20 \text{ A}, \text{ V}_{GS} = 0 \text{ V},$	-	72	_	nC
Reverse Recovery Time	t _{rr}	· R _G = 20 Ω, di/dt = 1000 A/μs, T _J = 25 °C	-	12	-	ns

ELECTRICAL CHARACTERISTICS (T_J = +25 $^{\circ}$ C unless otherwise specified) (continued)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
TYPICAL PERFORMANCE - REVERSE	DIODE	·		•		
Reverse Recovery Charge	Q _{rr}	$V_{DS} = 800 \text{ V}, \text{ I}_{S} = 20 \text{ A}, \text{ V}_{GS} = 0 \text{ V},$	_	110	-	nC
Reverse Recovery Time	t _{rr}	R _G = 20 Ω, di/dt = 1000 A/μs, T _J = 150 °C	-	23	-	ns
TYPICAL PERFORMANCE - DYNAMIC			U.			
Input Capacitance	C _{iss}	$V_{DS} = 800 \text{ V}, V_{GS} = 0 \text{ V},$	-	1370	-	pF
Output Capacitance	C _{oss}	f = 100 kHz	-	35	-	
Reverse Transfer Capacitance	C _{rss}		-	2	-	
Effective Output Capacitance, Energy Related	C _{oss(er)}	V_{DS} = 0 V to 800 V, V_{GS} = 0 V	-	42	-	pF
Effective Output Capacitance, Time Related	C _{oss(tr)}		-	71	-	pF
Coss Stored Energy	E _{oss}	$V_{DS} = 800 \text{ V}, \text{ V}_{GS} = 0 \text{ V}$	-	13.4	-	μJ
Total Gate Charge	Q _G	$V_{DS} = 800 \text{ V}, \text{ I}_{D} = 20 \text{ A},$	-	37.8	-	nC
Gate-drain Charge	Q _{GD}	V _{GS} = 0 V to 15 V	-	9.5	-	
Gate-source Charge	Q _{GS}		-	10	-	
Turn-on Delay Time	t _{d(on)}	$\begin{array}{l} V_{DS}=800 \text{ V}, \text{ I}_{D}=20 \text{ A},\\ \text{Gate Driver}=0 \text{ V to }+15 \text{ V},\\ \text{R}_{G,ON}=10 \ \Omega, \ \text{R}_{G,OFF}=50 \ \Omega,\\ \text{Inductive Load},\\ \text{FWD: Same Device With}\\ \text{V}_{GS}=0 \text{ V}, \ \text{R}_{G}=50 \ \Omega, \ \text{T}_{J}=25 \ ^{\circ}\text{C}\\ (\text{Note 6}) \end{array}$	-	39	-	ns
Rise Time	t _r		-	12	-	
Turn-off Delay Time	t _{d(off)}		-	155	-]
Fall Time	t _f		-	18	-	
Turn-on Energy	E _{ON}		-	437	-	μJ
Turn-off Energy	E _{OFF}		-	57	-	
Total Switching Energy	E _{TOTAL}		-	494	-	
Turn-on Delay Time	t _{d(on)}	$V_{DS} = 800 \text{ V}, \text{ I}_{D} = 20 \text{ A},$	-	38	-	ns
Rise Time	t _r	Gate Driver = 0 V to +15 V, R _{G.ON} = 10 Ω, R _{G.OFF} = 50 Ω,	-	14	-	
Turn-off Delay Time	t _{d(off)}	Inductive Load, FWD: Same Device With	-	162	-	
Fall Time	t _f	V_{GS} = 0 V, R_{G} = 50 Ω , T_{J} = 150 °C	-	21	-	
Turn-on Energy	E _{ON}	(Note 6)	-	540	-	μJ
Turn-off Energy	E _{OFF}		-	145	-	
Total Switching Energy	E _{TOTAL}		-	685	-	
Turn-on Delay Time	t _{d(on)}	$V_{DS} = 800 \text{ V}, I_D = 20 \text{ A},$	-	35	-	ns
Rise Time	t _r	Gate Driver = 0 V to +15 V, R _{G OFF} = 1 Ω , R _{G ON} = 5 Ω ,	-	7.2	-	
Turn-off Delay Time	t _{d(off)}	Snubber: $R_S = 10 \overline{\Omega}$, $C_S = 95 \text{ pF}$, Inductive Load,	-	32	-	
Fall Time	t _f	FWD: Same Device With	-	11	-	
Turn-on Energy Including R _S Energy	E _{ON}	V _{GS} = 0 V, R _G = 50 Ω, T _J = 25 °C (Notes 7, 8)	-	457	-	μJ
Turn-off Energy Including R _S Energy	E _{OFF}	1	-	75	-	
Total Switching Energy	E _{TOTAL}	1	-	532	-	
Snubber R_S Energy During Turn-on	E _{RS_ON}	1		5.6		
Snubber R _S Energy During Turn-off	E _{RS OFF}	1		5.7		

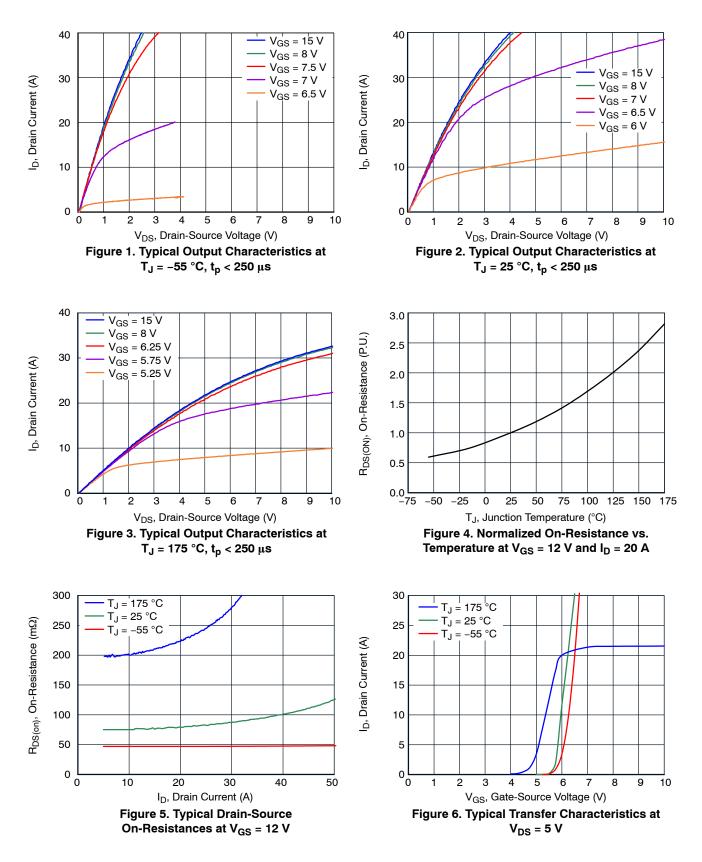
ELECTRICAL CHARACTERISTICS (T_J = +25 $^{\circ}$ C unless otherwise specified) (continued)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
TYPICAL PERFORMANCE - DYNAMIC						
Turn-on Delay Time	t _{d(on)}	$V_{DS} = 800 \text{ V}, I_D = 20 \text{ A},$	-	27	-	ns
Rise Time	t _r	Gate Driver = 0 V to +15 V, R _{G OFF} = 1 Ω , R _{G ON} = 5 Ω ,	-	14	-	
Turn-off Delay Time	t _{d(off)}	Snubber: $R_S = 10 \Omega$, $C_S = 95 pF$, Inductive Load,	-	33	-	
Fall Time	t _f	FWD: Same Device With $V_{GS} = 0 V, R_G = 50 \Omega$,	-	13	-	
Turn-on Energy Including R _S Energy	E _{ON}	$T_{\rm J} = 150 ^{\circ}{\rm C} ({\rm Notes}7,8)$	-	503	-	μJ
Turn-off Energy Including R _S Energy	E _{OFF}]	-	110	-	
Total Switching Energy	E _{TOTAL}		-	613	-	
Snubber R _S Energy During Turn-on	E _{RS_ON}]		5]
Snubber R _S Energy During Turn-off	E _{RS_OFF}]		5.1]

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product Product parametric performance is indicated in the Electrical Characteristics for the listed test condition performance may not be indicated by the Electrical Characteristics if operated under different conditions. 4. Limited by $T_{J,max}$. 5. Pulse width t_p limited by $T_{J,max}$. 6. Measured with the half-bridge mode switching test circuit in Figure 22. 7. Measured with the switching test circuit in Figure 23.

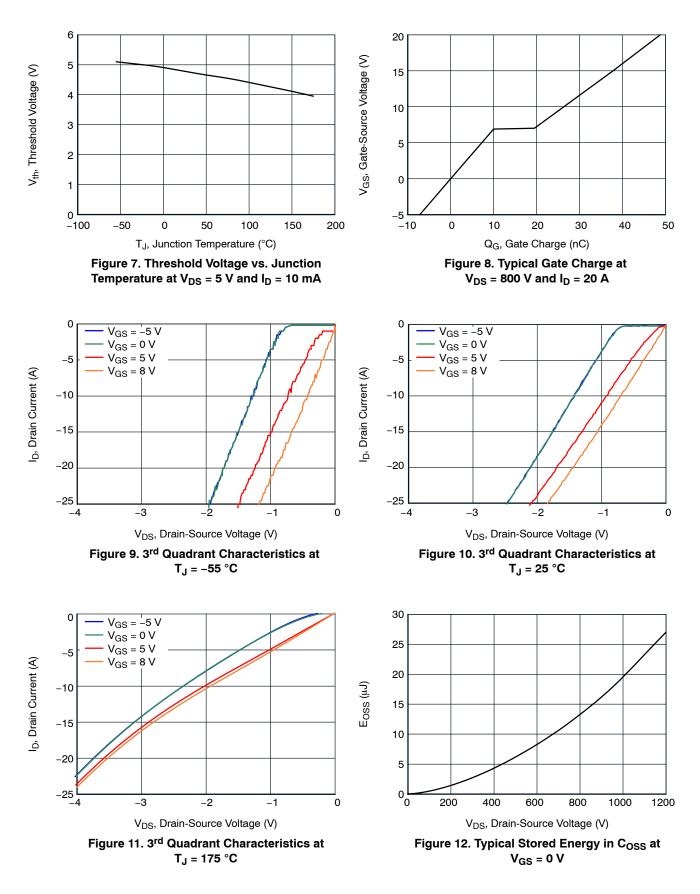
8. The switching energies (turn-on energy, turn-off energy and total energy) presented in this table include the device RC snubber energy losses.

TYPICAL PERFORMANCE DIAGRAMS

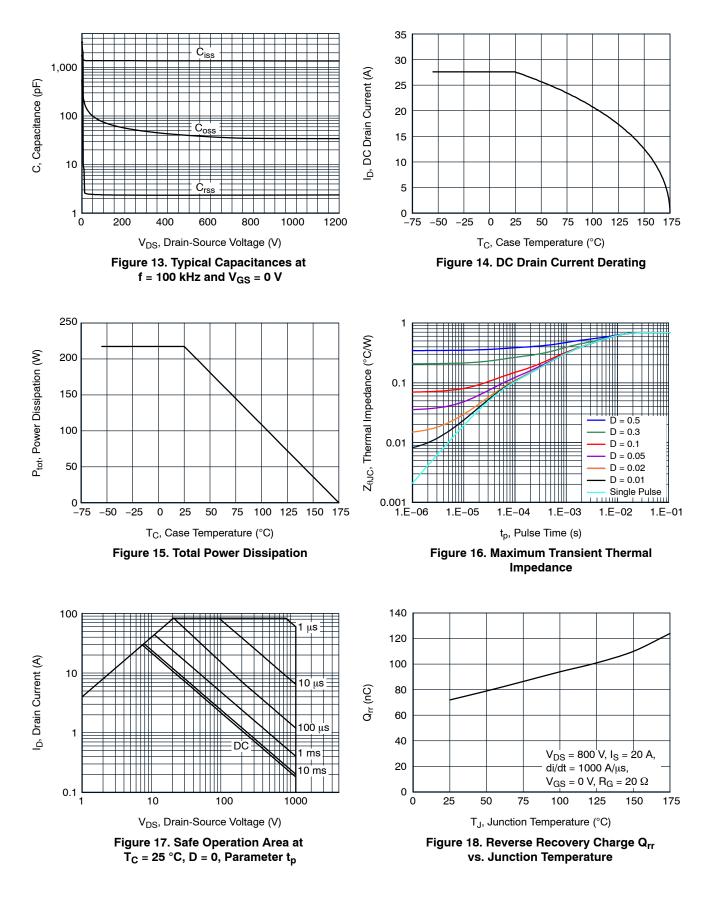


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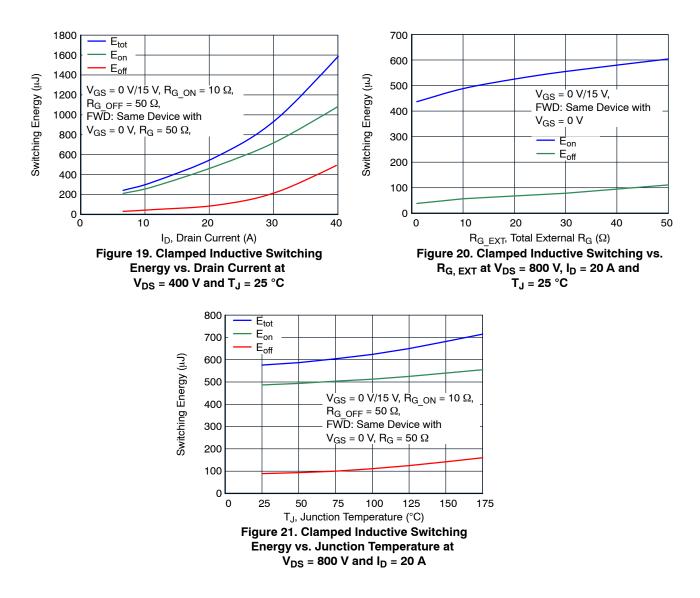
TYPICAL PERFORMANCE DIAGRAMS (continued)



TYPICAL PERFORMANCE DIAGRAMS (continued)



TYPICAL PERFORMANCE DIAGRAMS (continued)



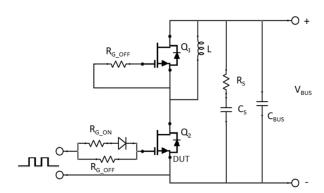


Figure 22. Schematic of the Half-Bridge Mode Switching Test Circuit. Note, a Bus RC Snubber ($R_S = 2.5 \Omega$, $C_S = 100 nF$) Is Used To Reduce the Power Loop High Frequency Oscillations

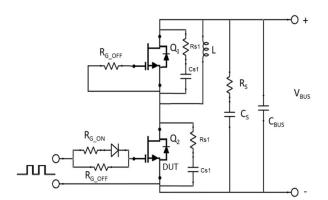


Figure 23. Schematic of the Half-Bridge Mode Switching Test Circuit With Device RC Snubbers (R_{S1} = 10 Ω , C_{S1} = 95 pF) and a Bus RC Snubber (R_S = 2.5 Ω , C_S = 100 nF)

APPLICATIONS INFORMATION

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (Q_G), and reverse recovery charge (Q_{rr}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

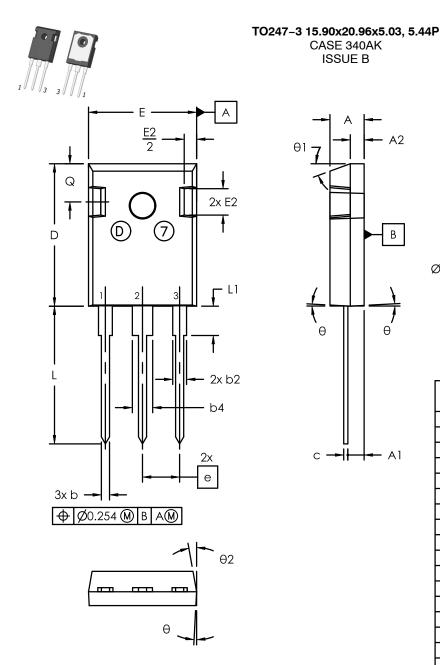
Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.onsemi.com.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small R(G) will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the **onsemi** website at www.onsemi.com.

ORDERING INFORMATION

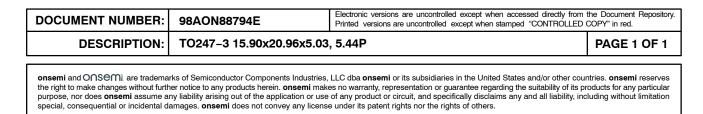
Part Number	Marking	Package	Shipping
UF4C120070K3S	UF4C120070K3S	TO247-3 15.90x20.96x5.03, 5.44P (Pb-Free, Halogen Free)	600 / Tube

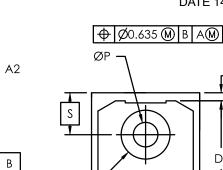
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NOTE:

- 1. Dimensioning and tolerancing as per ASME Y14.5 2018
- 2. Controlling dimension : millimeters
- 3. Package Outline in compliance with JEDEC standard var. AD.
- 4. Dimensions D & E does not include mold flash.
- ØP to have max draft angle of 1.7° to the top with max. hole 5. diameter of 3.91mm.





3

2

1

ØP1

θ

- A1

E1								
SYM	millimeters							
31101	MIN	NOM	MAX					
А	4.70	5.03	5.31					
A1	2.21	2.40	2.59					
A1 A2	1.50	2.03	2.49					
b	0.99	1.20	1.40					
b2	1.65	2.03	2.39					
b4 c D D1	2.59	3.00	3.43					
С	0.38	0.60	0.89					
D	20.70	20.96	21.46					
	13.08	-	-					
D2	0.51	1.19	1.35					
E	15.49	15.90	16.26					
е		5.44 BSC						
E1	13.00	13.30	13.60					
E2	3.43	3.89	5.20					
L	19.62	20.27	20.32					
L1	_	1	4.50					
ØP	3.40	3.60	3.80					
ØP1	7.06	7.19	7.39					
Q	5.38	5.62	6.20					
Q S		6.15 BSC						
θ		3°						
θ1	20°							
θ2	10°							

D2

Dl

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